This is the submitted version of a paper presented at *The 13th European Control Conference, 24th to the 27th of June 2014 in Strasbourg, France.*

Citation for the original published paper:


N.B. When citing this work, cite the original published paper.

Permanent link to this version:
http://urn.kb.se/resolve?urn=urn:nbn:se:kth:diva-136301
Discrete-time Network-based Control under Try-Once-Discard Protocol and Actuator Constraints

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Abstract—This paper deals with the solution bounds for discrete-time Networked Control Systems (NCSs) via delay-dependent Lyapunov-Krasovskii methods. Solution bounds are widely used for systems with input saturation caused by actuator saturation or by the quantizers with saturation. A time-delay approach was introduced recently for continuous-time NCS under a weighted Try-Once-Discard (TOD) protocol in [6], where actuator saturation was not taken into account. In the present paper, we develop the time-delay approach for linear (probably, uncertain) discrete-time NCS under the weighted TOD protocol in the presence of actuator saturation. A novel Lyapunov-based method is presented for finding the domain of attraction. Polytopic uncertainties in the system model can be easily included in our analysis. The efficiency of the time-delay approach is illustrated on the example of a cart-pendulum system.

Keywords: Networked control systems, time-delay approach, scheduling, input saturation, Lyapunov-Krasovskii method.

I. INTRODUCTION

Networked Control Systems are control systems comprised of the system to be controlled and of actuators, sensors, and controllers, the operation of which is coordinated via a communication network [1]. In the present paper, we focus on the stability analysis of discrete-time NCS with communication constraints and actuator constraints. A linear (probably, uncertain) system with distributed sensors is considered.

The time-delay approach has been introduced recently for the stabilization of continuous-time NCS under the Round-Robin (RR) protocol in [7] and under a weighted Try-Once-Discard (TOD) protocol in [6], respectively. The closed-loop system is modeled as a switched system with multiple and ordered time-varying delays under RR scheduling or as a hybrid system with time-varying delays in the dynamics and in the reset equations under the TOD scheduling. Differently from the existing hybrid [4] and discrete-time [3] approaches, the transmission delay is allowed to be larger than the sampling interval. However, actuator saturation has not been taken into account in [6] and [7]. Recently, the stabilization of sampled-data systems under variable samplings and actuator saturation was studied in [10], where scheduling protocols and delays are not included.

As was pointed out in [5], when we deal with the solution bounds of time-delay systems via Lyapunov-Krasovskii method, the first time-interval of the delay length needs a special analysis. Solution bounds are widely used for systems with input saturation caused by actuator saturation or by the quantizers with saturation. This first time-interval does not influence on the stability and the exponential decay rate analysis. The analysis of the first time-interval is important for nonlinear systems e.g., for finding the domain of attraction.

In the present paper, the time-delay approach is extended to the stability analysis of discrete-time NCS under the weighted TOD protocol, actuator saturation, variable sampling intervals and variable delays. Following [5], we present a Lyapunov-based method for finding the domain of attraction under the weighted TOD protocol. The conditions are given in terms of Linear Matrix Inequalities (LMIs). Polytopic uncertainties in the system model can be easily included in the analysis. The efficiency of the presented approach is illustrated by a cart-pendulum system.

Notation: Throughout the paper the superscript ‘$T$’ stands for matrix transposition, $\mathbb{R}^n$ denotes the $n$ dimensional Euclidean space with vector norm $| \cdot |$, $\mathbb{R}^{n \times m}$ is the set of all $n \times m$ real matrices, and the notation $P > 0$, for $P \in \mathbb{R}^{n \times n}$ means that $P$ is symmetric and positive definite. The symmetric elements of the symmetric matrix will be denoted by $\ast$, $\mathbb{Z}$ and $\mathbb{N}$ denote the set of non-negative integers and positive integers, respectively. For any matrix $A \in \mathbb{R}^{n \times n}$ and vector $x \in \mathbb{R}^n$, the notations $A_i$ and $x_i$ denote, respectively, the $j^{th}$ line of matrix $A$ and the $j^{th}$ component of vector $x$. $\mathbb{Z}$ denotes the set of non-negative integers. Given $\bar{u} = [\bar{u}_1, \ldots, \bar{u}_n]^T$, $0 < \bar{u}_i$, $i = 1, \ldots, n$, for any $u = [u_1, \ldots, u_n]^T$, we denote by $sat(u, \bar{u})$ the vector with coordinates $\text{sign}(u_i)\min(|u_i|, \bar{u}_i)$.

II. SYSTEM MODEL

A. Discrete-time NCS model

Consider the system architecture in Fig.1 with discrete-time plant:

$$x(t+1) = Ax(t) + Bu(t), \quad t \in \mathbb{Z},$$

(1)

where $x(t) \in \mathbb{R}^n$ is the state vector, $u(t) \in \mathbb{R}^{n_u}$ is the control input. $A$ and $B$ are (probably, uncertain) system matrices of appropriate dimensions. The initial condition is $x(0) = x_0$. We suppose that the control input is subject to the following amplitude constraints

$$|u_i(t)| \leq \bar{u}_i, \quad 0 < \bar{u}_i, \quad i = 1, \ldots, n_u, \quad t \in \mathbb{Z}.$$

(2)
The NCS has several nodes (distributed sensors, a controller node and an actuator node) which are connected via networks. For the sake of simplicity, we consider two sensor nodes $y_i(t) = C_1x(t), i = 1, 2$ and we denote $C = [c_1^T, c_2^T]^T$, $y(t) = \left[\begin{array}{c} y_1(t) \\ y_2(t) \end{array} \right] \in \mathbb{R}^{n_y}$. The results can be easily extended to any finite number of sensors. We let $s_k$ denote the unbounded and monotonously increasing sequence of sampling instants

$$0 = s_0 < s_1 < \ldots < s_k < \ldots, \quad k \in \mathbb{Z}, \lim_{k \to \infty} s_k = \infty, \quad (3)$$

where $\{s_k\}$ is a subsequence of $\mathbb{Z}$. At each sampling instant $s_k$, one of the outputs $y_i(t) \in \mathbb{R}^{n_i}(n_1 + n_2 = n_y)$ is sampled and transmitted over the network.

![Fig. 1. NCS with actuator saturation under TOD protocol](image)

It is assumed that no packet dropouts and no packet disorders will happen during the data transmission over the network. The transmission of the information over the two networks (between the sensor and the actuator) is subject to a variable total delay $\eta_k = \eta_k^c + \eta_k^a$, where $\eta_k^c \in \mathbb{Z}$ and $\eta_k^a \in \mathbb{Z}$ are the network-induced delays from the sensor to the controller and from the controller to the actuator, respectively. Then $t_k = s_k + \eta_k$ is the updating time instant of the Zero-Order Hold (ZOH), where $\{t_k\}$ is a subsequence of $\mathbb{Z}$.

Following [7] and [8], we allow the delay to be non-small provided that the old sample cannot get to the destination (to the controller or to the actuator) after the most recent one. Assume that the network-induced delay $\eta_k$ and the time span between the updating and the most recent sampling instant are bounded:

$$t_{k+1} - t_k + \eta_k \leq 1 + \tau_M, \quad \text{i.e.} \quad (t_{k+1} - 1) - s_k \leq \tau_M, \quad 0 \leq \eta_m \leq \eta_k \leq \eta_m, \quad k \in \mathbb{Z}. \quad (4)$$

Here $\tau_M, \eta_m$, and $\eta_M$ are known non-negative integers and $\tau_M = MATT + \eta_M$, where $MATT$ denotes the Maximum Allowable Transmission Interval. Then

$$(t_{k+1} - 1) - t_k \leq \tau_M - \eta_m,$$

$$(t_{k+1} - 1) - t_k + \eta_k \leq 2\tau_M - \eta_m + 1 \leq \tau_M.$$  \quad (5)

Note that the first updating time $t_0$ corresponds to the first data received by the actuator, which means that $u(t) = 0, t \in [0, t_0 - 1]$. Then for $t \in [0, t_0 - 1]$, (1) is given by

$$x(t + 1) = Ax(t), \quad t = 0, 1, \ldots, t_0 - 1, \quad t \in \mathbb{Z}. \quad (6)$$

### B. TOD protocol and a discrete-time hybrid time-delay model

In [6], a weighted TOD protocol was introduced for the stabilization of continuous-time NCS in terms of time-delay approach. Actuator saturation was not taken into account. Following [6], consider the error between the system output $y(s_k)$ and the last available information $\hat{y}(s_k-1)$:

$$e(t) = col\{e_1(t), e_2(t)\} \equiv \hat{y}(s_{k-1}) - y(s_k), \quad t \in [t_k, t_{k+1} - 1], \quad t \in \mathbb{Z}, \quad k \in \mathbb{Z}, \quad \hat{y}(s_{-1}) \equiv 0, \quad e(t) \in \mathbb{R}^{n_y}.$$  

Let $Q_1 > 0, i = 1, 2$ be some weighting matrices (they will be found from LMIs in Lemma 2 below). The node that has the largest error, $|\sqrt{Q_1}e_1(t)|^2 < i = 1, 2$ is granted access to the network. Consider the first case of

$$|\sqrt{Q}_1e_1(t)|^2 \geq |\sqrt{Q}_2e_2(t)|^2, \quad t \in [t_k, t_{k+1} - 1], \quad t \in \mathbb{Z}. \quad (7)$$

Due to the control bounds defined in (2), the effective control signal to be applied to the system (1) is given by

$$u(t) = \text{sat}[K_1y(t - \eta_k) + K_2\hat{y}(t - 1 - \eta_k)], \quad t \in [t_k, t_{k+1} - 1], \quad t \in \mathbb{Z}.$$  

We define the polyhedron

$$\mathcal{L}(K^j, \hat{u}) = \{x \in \mathbb{R}^n : \langle \langle K^jC^j \rangle \rangle, x \rangle \leq \frac{1}{2} \hat{u}_i, \quad i = 1, \ldots, n_u\},$$

where $j = 1, 2$. If the control is such that $x \in \mathcal{L}(K^1, \hat{u}) \cap \mathcal{L}(K^2, \hat{u})$, then $\langle \langle K^1C^1 \rangle \rangle, x + \langle \langle K^2C^2 \rangle \rangle, x \rangle \leq \hat{u}$. Following [6], we obtain thus the impulsive closed-loop model with the following discrete dynamics:

$$x(t + 1) = Ax(t) + A_1x(t_k - \eta_k) + B_2e_2(t), \quad e(t + 1) = e(t), \quad t \in [t_k, t_{k+1} - 1], \quad t \in \mathbb{Z},\quad (8)$$

where $A_1 = BK_1$, $B_2 = BK_2$, $K = [K_1 K_2], \quad i = 1, 2$. The delayed reset system is given by

$$x(t_{k+1}) = x(t_{k+1}^0), \quad e(t_{k+1}) = \left[\begin{array}{c} e_1(t_{k+1}) \\ e_2(t_{k+1}) \end{array} \right] + Cx(t_k - \eta_k) - Cx(t_{k+1} - \eta_{k+1}). \quad (9)$$

Similarly if

$$|\sqrt{Q}_2e_2(t)|^2 > |\sqrt{Q}_1e_1(t)|^2, \quad t \in [t_k, t_{k+1} - 1], \quad t \in \mathbb{Z},$$

then the closed-loop system has the form

$$x(t + 1) = Ax(t) + A_1x(t_k - \eta_k) + B_1e_1(t), \quad e(t + 1) = e(t), \quad t \in [t_k, t_{k+1} - 1], \quad t \in \mathbb{Z}, \quad (11)$$

with the delayed reset system

$$x(t_{k+1}) = x(t_{k+1}^0), \quad e(t_{k+1}) = \left[\begin{array}{c} e_1(t_{k+1}) \\ e_2(t_{k+1}) \end{array} \right] + Cx(t_k - \eta_k) - Cx(t_{k+1} - \eta_{k+1}). \quad (12)$$
where $A_1 = BK$, $B_1 = BK^i$, $i = 1, 2$. The initial condition for (7)-(12) has the form of (6) and $e(t_0) = -C x(t_0 - \eta_0) = -C x_0$. 

Denote by $x(t, x_0)$ the state trajectory of (7)-(12) with respect to $x$ and with the initial condition $x_0 \in \mathbb{R}^n$. Then the domain of attraction of the origin of the closed-loop system (7)-(12) is the set $A = \{ x_0 \in \mathbb{R}^n : \lim_{t \to \infty} x(t, x_0) = 0 \}$. Given $K^i, K^2$ and non-negative integers $0 \leq \eta_m \leq \tau_m < \tau_M$, our objective is to get $A_\beta \subset A$ (as large as we can get) on the domain of attraction, for which exponential stability of the closed-loop system (7)-(12) with respect to $x$ is ensured, where

$$A_\beta = \{ x_0 \in \mathbb{R}^n : x_0^T P x_0 \leq \beta^{-1} \},$$

and where $\beta > 0$ is a scalar, $P > 0$ is an $n \times n$-matrix.

### III. Partial Exponential Stability of the Hybrid Delayed System Without Actuator Saturation

**Definition 1:** Hybrid system (7)-(12) with initial conditions (6) and $e(t_0)$ is said to be partially exponential stability with respect to $x$ if there exist constants $b > 0$, $0 < \kappa < 1$ such that the following holds

$$|x(t)|^2 \leq b e^{-\kappa t} \left[ |x_0|^2 + |e(t_0)|^2 \right], \quad t \geq t_0$$

for the solutions of the hybrid system initialized with (6) and $e(t_0) \in \mathbb{R}^n$.

We extend the result here of [6] to the discrete-time case. Consider the Lyapunov-Krasovskii Functional (LKF) of the form:

$$V_c(t) = V_{TOD}(t) + \min_{i=1,2} \{ e_i^T(t) Q_i e_i(t) \},$$

where

$$V_{TOD}(t) = V(t) = (\tau_M - \eta_m) \sum_{k=t_0-k}^{t_0-k-1} \lambda^{k-1} s^T(s) Q(s) s,$$

and along (10)-(12)

$$V_c(t + 1) - \lambda V_c(t) - \frac{1}{\tau_M - \eta_m} \| \sqrt{Q} e_1(t) \|^2 \leq 0,$$

hold with

$$Q \lambda^T \geq C^T Q C, \quad i = 1, 2.$$  

Then $V_c(t)$ does not grow at the jumps along (7)-(9), where

$$V_c(t_{k+1}) - V_c(t_k) \leq -\| \sqrt{Q} e_1(t_k) \|^2, \quad i = 2,$$

and along (10)-(12), where (20) holds with $i = 1$. Moreover, the following bound is valid for the solution of (7)-(12) with the initial condition (6) and $e(t_0) \in \mathbb{R}^n$:

$$V_{TOD}(t) \leq \lambda^{t_0} V_c(t_0), \quad t \geq t_0, \quad t \in \mathbb{Z},$$

yielding exponential stability of (7)-(12) with respect to $x$.

**Proof:** Consider first the case of (7), where

$$V_c(t) = V_{TOD}(t) + \| \sqrt{Q} e_2(t) \|^2, \quad t \in [t_k, t_{k+1} - 1].$$

Since $\sum_{s=t_0-1}^{t_0-1} \lambda^{s-1} \leq \tau_M - \eta_m$, $t \in (t_k, t_{k+1} - 1]$, by the comparison principle, (17) implies

$$V_c(t) \leq \lambda^{t_0-t_k} V_c(t_k) + \| \sqrt{Q} e_2(t_k) \|^2, \quad t \in [t_k, t_{k+1} - 1].$$

Therefore,

$$V_{TOD}(t) \leq \lambda^{t_0-t_k} V_c(t_k), \quad t \in [t_k, t_{k+1} - 1].$$

If

$$\| \sqrt{Q} e_1(t_k) \|^2 - \| \sqrt{Q} e_2(t_k) \|^2 \geq 0$$

then

$$V_c(t_{k+1}) = V(t_k) + \| \sqrt{Q} e_2(t_{k+1}) \|^2 + (\tau_M - \eta_m) \sum_{s=t_k-k}^{t_k-k-1} \lambda^{s-1} \eta^T(s) Q(s) \eta(s),$$

and due to (9)

$$\| \sqrt{Q} e_1(t_{k+1}) \|^2 = \| \sqrt{Q} C_1 [x(t_k - \eta_k) - x(t_k - \eta_{k+1})] \|^2.$$  

Denote $\Theta \triangleq V_c(t_{k+1}) - V_c(t_k) + \| \sqrt{Q} e_2(t_{k+1}) \|^2$. Taking into account (16), we obtain

$$\Theta \leq \| \sqrt{Q} e_1(t_{k+1}) \|^2 - \lambda^{\tau_M} \| \sqrt{Q} [x(t_k - \eta_k) - x(t_k - \eta_{k+1})] \|^2.$$  

Due to (24), we have further

$$\Theta \leq \| \sqrt{Q} e_1(t_{k+1}) \|^2 - \lambda^{\tau_M} \| \sqrt{Q} [x(t_k - \eta_k) - x(t_k - \eta_{k+1})] \|^2 \leq [x(t_k - \eta_k) - x(t_k - \eta_{k+1})]^T (Q \lambda^T - C_q C_q^T) [x(t_k - \eta_k) - x(t_k - \eta_{k+1})] \leq 0,$$

where the latter inequality follows from (19) with $i = 1$. If

$$\| \sqrt{Q} e_2(t_{k+1}) \|^2 > \| \sqrt{Q} e_1(t_{k+1}) \|^2$$

then

$$V_c(t_{k+1}) = V(t_{k+1}) + \| \sqrt{Q} e_1(t_{k+1}) \|^2 + (\tau_M - \eta_m) \sum_{s=t_k-k}^{t_k-k-1} \lambda^{s-1} \eta^T(s) Q(s).$$
and similarly
\[ \Theta \le (|\sqrt{Q_1}e_1(t_{k+1})|^2 - |\sqrt{Q_2}e_2(t_k)|^2) + |\sqrt{Q_2}e_2(t_k)|^2 \]
\[ -\lambda^{\tau_m} |\sqrt{Q_1}\{x(t_k - \eta_k) - x(t_{k+1} - \eta_{k+1})\}|^2 \]
\[ \le -(x(t_m - \eta_m) - x(t_{m+1} - \eta_{m+1}))^T [Q\lambda^{\tau_m} - C^TQ_1C] \]
\[ \times [x(t_k - \eta_k) - x(t_{k+1} - \eta_{k+1})] \le 0. \]
Therefore, (20) with \( i = 2 \) is valid and, together with (22)
for \( t = t_{k+1} \), implies
\[ V_x(t_{k+1}) \le \lambda^{\tau_m - \tau_{k+1}} V_x(t_k). \]
(25)
Similarly, if \( |\sqrt{Q_2}e_2(t_k)|^2 > |\sqrt{Q_1}e_1(t_k)|^2 \), then (18) and (19) with \( i = 2 \) imply (23), (20) with \( i = 1 \) and (25). Hence, (25) holds for all \( k \in \mathbb{Z} \)
and
\[ V_x(t_{k+1}) \le \lambda^{\tau_m - \tau_{k+1}} V_x(t_{k+1}). \]
(26)
The inequality with \( k + 1 \) replaced by \( k \) and (23) imply (21) that completes the proof. \( \square \)

Based on the above lemma, by using the standard arguments for the delay-dependent analysis [9], we derive LMI
conditions for the exponential stability with respect to \( x \):

**Lemma 2**: Given scalar \( \lambda < 1 \), non-negative integers
\( 0 \leq \eta_m \leq \eta_M < \tau_m \), and \( K^1, K^2 \). If there exist \( n \times n \) matrices \( P > 0 \), \( Q > 0 \), \( S \), \( R_i > 0 \) \( i = 0, 1 \), \( S_{12} \), matrices \( Q_1 > 0 \), \( Q_2 > 0 \) such that (19) and
\[ \Omega = \begin{bmatrix} R_1 & S_{12} & S_{12} & R_1 \\ \end{bmatrix} \geq 0, \]
(27)
are feasible, where
\[ F_0 = [A 0 0 1 0 B_{3-i}], \quad F_j = [\Gamma_1 \Gamma_2 \Gamma_3 \Gamma_4 \Gamma_5], \]
\[ \Gamma_j = I, \Gamma_l = 0, l = 1, \ldots, 5, l \neq j, \]
\[ F_{k,j} = F_k - F_j, \quad k, j = 0, 1, \ldots, 5, \]
\[ F_{01} = F_0 - F_1, \quad F = [F_{23} F_{31} F_{34}^T], \]
\[ \Sigma = diag\{S_0 - \lambda P - (S_0 - S_1)\lambda^{\eta_m}, 0, -S_1\lambda^{\tau_m}, \varphi\}, \]
\[ \varphi = [-1 - \lambda^{\tau_m - \tau_{k+1}} (1 - \lambda)]Q_{3-i}, \]
\[ W = Q_{m}^T R_{0} + (\tau_M - \eta_m)^2 R_1 + (\tau_M - \eta_m)Q, \quad i = 1, 2. \]
(28)

Then solutions of the hybrid system (7)-(12) satisfy the bound (21) and are exponentially stable with respect to \( x \).

**IV. PARTIAL EXPONENTIAL STABILITY OF THE HYBRID DELAYED SYSTEM WITH ACTUATOR SATURATION**

**A. UNDER THE CONSTANT INITIAL CONDITION (15) AND V_{TOD}(t) OF (14)**

For \( t = 0 \)
\[ V_{TOD}(0) = x_{TOD}^T P x_0 + \sum_{s=1}^{n-1} \eta_m \lambda^{-s} x_0^T S_0 x_0 + \sum_{s=1}^{n-1} \lambda^{-s} x_0^T T_1 S_1 x_0. \]
(29a)
Thus,
\[ V_{TOD}(0) \leq x_{TOD}^T \Xi_{TOD} x_0, \]
(29b)
where
\[ \Xi_{TOD} = P + \eta_m S_0 + \lambda^{\eta_m} (\tau_M - \eta_m) S_1. \]
In order to derive the bound on \( V_{TOD}(t) \) in terms of \( x_0 \), by the arguments in [5] we consider \( V_0(t) = x^T(t) P x(t), \)
\( P > 0 \) and have for \( t = 0, 1, \ldots, t_0 - 1 \):
\[ V_0(t+1) - \mu V_0(t) \leq 0, \quad \mu > 1, \]
(29a)
\[ V_{TOD}(t+1) - \mu V_{TOD}(t) - (\mu - 1) V_0(t) \leq 0. \]
(29b)

Then from (29a), \( V_0(t) \leq \mu^t V_0(0) \) for \( t = 0, 1, \ldots, t_0 \).
Noting that \( \eta_m \leq \eta_0 = 0 \leq \eta_M \), we have from (29b)
\[ V_{TOD}(t_0) \leq \lambda^{t_0} V_{TOD}(0) + (\mu^{t_0} - 1) x_0^T P x_0 \]
\[ \leq x_0^T \lambda^{t_0} \Xi_{TOD} + (\mu^{t_0} - 1) P x_0 \]
\[ \leq x_0^T \lambda^{t_0} \Xi_{TOD} + (\mu^{t_{M-1}} - 1) P x_0. \]
(30)

**Theorem 1**: Given positive scalars \( \lambda < 1, \beta \geq 1, \sigma, \beta \),
non-negative integers \( 0 \leq \eta_m \leq \eta_M < \tau_m \), and \( K^1, K^2 \). Suppose there exist \( n \times n \) matrices \( P > 0 \), \( Q > 0 \), \( S \), \( R_i > 0 \) \( i = 0, 1 \), \( S_{12} \), matrices \( Q_1 > 0 \), \( Q_2 > 0 \) such that
(19), (26) and
\[ \left[ \begin{array}{cc} -\mu P & AT^P \\ A^T P & -P \end{array} \right] \geq 0, \]
(31)
\[ \eta_0 S_0 + \lambda^{\eta_m} (\tau_M - \eta_m) S_1 \leq \sigma P, \]
(32)
\[ Q \lambda^{t_M} \leq \sigma P, \]
(33)
Then, for all initial conditions \( x_0 \) belonging to \( X_0 \), the closed-loop system (7)-(12) is exponentially stable with respect to \( x \).

**Proof**: Suppose that \( x(t) \in \mathcal{L}(K^1, \bar{u}) \cap \mathcal{L}(K^2, \bar{u}) \). As was shown in Lemma 2, LMIs (19), (26) and (27) lead to (21) for \( t = t_0, t_0 + 1, \ldots \). By applying convex analysis of [9], we show that for \( t = 0, 1, \ldots, t_0 - 1 \), LMIs (31) and (35) with (26) can guarantee (29a) and (29b), respectively.

Next, noting that (19) and (33), we have
\[ x_0^T C^{T} Q_1 C x_0 \leq \lambda^{t_M} x_0^T Q x_0 \leq \sigma x_0^T P x_0, \]
(36)
which implies that \( \left| \sqrt{Q_1} e_i(t_0) \right|^2 = | - \sqrt{Q_1} C x_0 |^2 < \sigma x_0^T P x_0, \quad i = 1, 2 \).

Therefore, taking into (21), (30), (32), (36) and \( t_0 = \eta_0 \leq \eta_M \), we obtain
\[ x^T(t) P x(t) \leq V_{TOD}(t) \leq \lambda^{t_0} V_{TOD}(t) \]
\[ = \lambda^{t_0} \left( V_{TOD}(t_0) + \min_{i=1,2} \{e_i^T(t_0) Q(T_0) e_i(t_0)\} \right) \]
\[ \leq \lambda^{t_0} x_0^T \left( \lambda^{t_0} \Xi_{TOD} + \lambda^{t_0} - 1 P + \sigma P \right) x_0 \]
\[ \leq \lambda^{t_0} x_0^T \left( \lambda^{t_0} (1 + \sigma) + (\mu^{t_{M-1}} - 1) + \sigma P \right) x_0 \]
\[ = \lambda^{t_0} \rho \tau_{TOD} x_0^T P x_0 \leq \lambda^{t_0} \rho \tau_{TOD} x_0^T P x_0, \quad t \in \mathbb{Z}. \]

So for all \( x(t) \in \mathcal{L}(K^1, \bar{u}) \cap \mathcal{L}(K^2, \bar{u}) \),
\[ x^T(t) (K^1 C_i^T)^T (K^2 C_i^T) x(t) \leq \frac{1}{\beta} a_{11}^2, \quad i = 1, \ldots, n_u, \quad j = 1, 2. \]
The latter inequality is guaranteed if 
\( \frac{1}{2} \beta \lambda^{\text{min}} \rho_{\text{TOD}}^{-1} P \bar{u}_2^2 - (K^T C_1)^T (K^T C_1) \geq 0 \), and, thus, by Schur complements if (34) is feasible. Hence, the solutions of hybrid system (7)-(12) with respect to \( x \) converge to the origin exponentially, provided that \( x_0 \in X_0 \).

Remark 1: Note that
\[
 x_0^T P x_0 \leq \lambda_{\text{max}}(P) |x_0|^2 < \beta^{-1},
\]
where \( \lambda_{\text{max}}(P) \) denotes the largest eigenvalue of \( P \). Hence the following initial region \( |x_0|^2 < \beta^{-1}/\lambda_{\text{max}}(P) \) is inside of \( X_0 \). In order to maximize the initial ball, we can add the condition
\[
 P - \gamma I < 0,
\]
to Theorem 1, where \( \gamma > 0 \) is minimized.

V. EXAMPLE: DISCRETE-TIME CART-PENDULUM

Consider the following linearized model of the inverted pendulum on a cart [7]:
\[
\begin{bmatrix}
\dot{x}(t) \\
\dot{\theta}(t)
\end{bmatrix}
= \begin{bmatrix}
x(t) \\
\theta(t)
\end{bmatrix} + K \begin{bmatrix}
0 \\
1
\end{bmatrix} u(t)
\]
with \( M = 3.9249\text{Kg}, m = 0.2047\text{Kg}, l = 0.2302\text{m}, g = 9.81\text{N/Kg}, a = 25.3\text{N/V} \) and \( \bar{u} = 50 \). In the model, \( x \) and \( \theta \) represent cart position coordinate and pendulum angle from vertical, respectively. Such a model is discretized with a sampling time \( T_s = 0.001s \):
\[
\begin{bmatrix}
\Delta x(t + 1) \\
\Delta \theta(t + 1)
\end{bmatrix}
= \begin{bmatrix}
1 & 0.001 & 0 & 0 \\
0 & 0 & 1 & 0.001
\end{bmatrix} \begin{bmatrix}
x(t) \\
\theta(t)
\end{bmatrix}
+ \begin{bmatrix}
0 \\
0.0064 \\
-0.0280
\end{bmatrix} u(t), \ t \in \mathbb{Z}
\]
with \( \bar{u} = 50 \). The pendulum can be stabilized by a state feedback \( u(t) = K \begin{bmatrix} x & \dot{x} & \theta & \dot{\theta} \end{bmatrix}^T \) with the gain \( K = [K^1, K^2] \)
\[
K^1 = \begin{bmatrix}
5.825 \\
5.883
\end{bmatrix}, K^2 = \begin{bmatrix}
24.941 \\
5.140
\end{bmatrix},
\]
which leads to the closed-loop system eigenvalues \( \{0.8997, 0.9980 + 0.0020i, 0.9980 - 0.0020i, 0.9980\} \). Suppose the variables \( \theta, \Delta \theta \) and \( x, \Delta x \) are not accessible simultaneously. We consider measurements \( y^i(t) = C^i x(t), \ t \in \mathbb{Z} \), where
\[
C^1 = \begin{bmatrix}
1 & 0 & 0 & 0 \\
0 & 1 & 0 & 0
\end{bmatrix}, \ C^2 = \begin{bmatrix}
0 & 0 & 1 & 0 \\
0 & 0 & 0 & 1
\end{bmatrix}.
\]
Choose \( \lambda = 1, \ \beta = 1, \ \sigma = 1.0 \times 10^{-2} \) and \( \eta_m = 1, \ \eta_M = 2, \ \tau_M = 3 \). Applying Theorem 1 with \( \mu = 1.02, \ \sigma = 1.1 \) and Remark 1, the closed-loop system (7)-(12) is exponentially stable with respect to \( x \) starting from the initial ball \( |x_0| < 0.3342 \) by weighted TOD protocol. Then for different \( \mu \), applying Theorem 1 with \( \lambda = 1 \) and Remark 1, we give the corresponding largest ball of admissible initial conditions in Table 1, from which we can see that \( |x_0| \) increases with smaller \( \mu \).

<table>
<thead>
<tr>
<th>\mu \</th>
<th>1.02</th>
<th>1.2</th>
<th>3</th>
</tr>
</thead>
<tbody>
<tr>
<td>(</td>
<td>x_0</td>
<td>) (Theorem 1, TOD)</td>
<td>0.3342</td>
</tr>
</tbody>
</table>

VI. CONCLUSIONS

In this paper, a time-delay approach was developed to the stability analysis of discrete-time NCS under a weighted TOD protocol in the presence of actuator saturation. A Lyapunov-based method was presented for finding the domain of attraction under the weighted TOD protocol. The conditions are given in terms of LMIs. Polytopic uncertainties in the system model can be easily included in the analysis. Numerical example illustrates the efficiency of our method.

REFERENCES